

CE 513: STATISTICAL METHODS IN CIVIL ENGINEERING

Lecture: Introduction to Fourier transforms

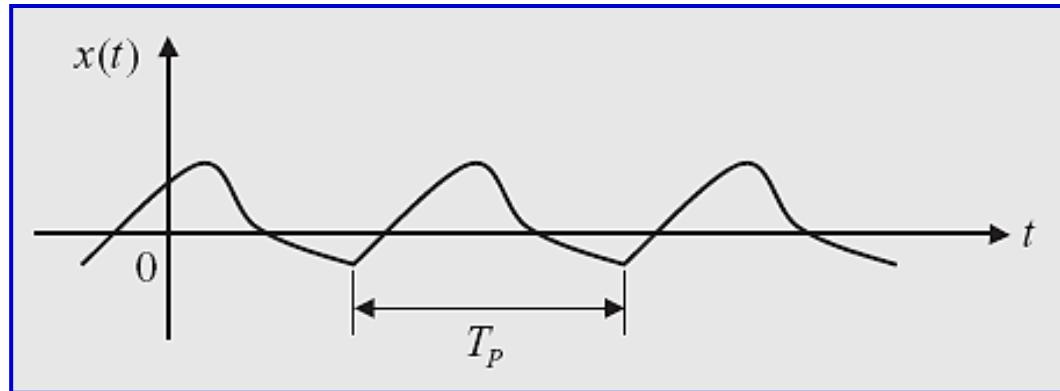
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Room: N-307

Department of Civil Engineering



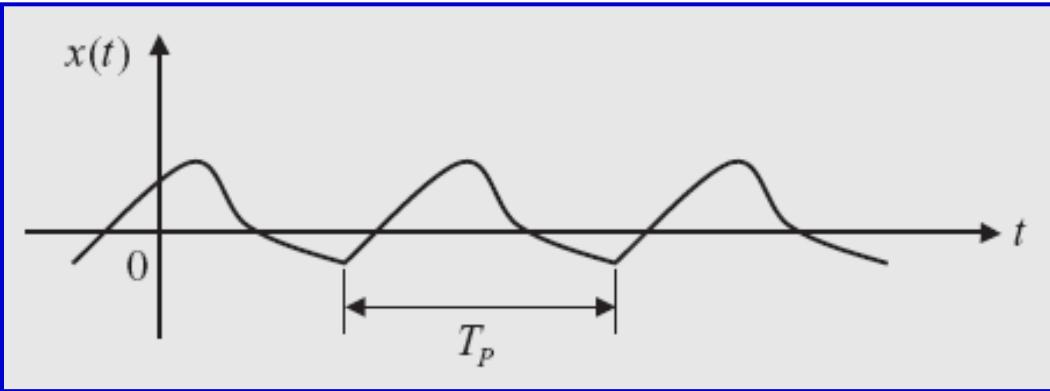
Fourier Analysis



A period signal with a period T_P

$$x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} \left[a_n \cos\left(\frac{2\pi n t}{T_P}\right) + b_n \sin\left(\frac{2\pi n t}{T_P}\right) \right]$$

Fourier Series



$$\frac{a_0}{2} = \frac{1}{T_P} \int_0^{T_P} x(t) dt = \frac{1}{T_P} \int_{-T_P/2}^{T_P/2} x(t) dt : \text{ mean value}$$

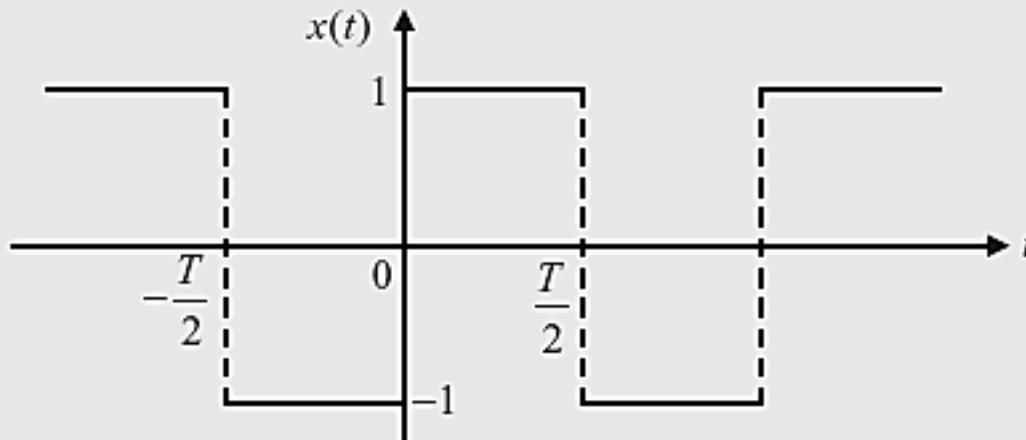
$$a_n = \frac{2}{T_P} \int_0^{T_P} x(t) \cos\left(\frac{2\pi nt}{T_P}\right) dt = \frac{2}{T_P} \int_{-T_P/2}^{T_P/2} x(t) \cos\left(\frac{2\pi nt}{T_P}\right) dt \quad n = 1, 2, \dots$$

$$b_n = \frac{2}{T_P} \int_0^{T_P} x(t) \sin\left(\frac{2\pi nt}{T_P}\right) dt = \frac{2}{T_P} \int_{-T_P/2}^{T_P/2} x(t) \sin\left(\frac{2\pi nt}{T_P}\right) dt \quad n = 1, 2, \dots$$

Fourier Series: Example

$$x(t) = \begin{cases} -1 & -\frac{T}{2} < t < 0 \\ 1 & 0 < t < \frac{T}{2} \end{cases}$$

and $x(t + nT) = x(t) \quad n = \pm 1, \pm 2, \dots$



Fourier Series: Example

$$a_0 = 0; \quad a_n = 0$$

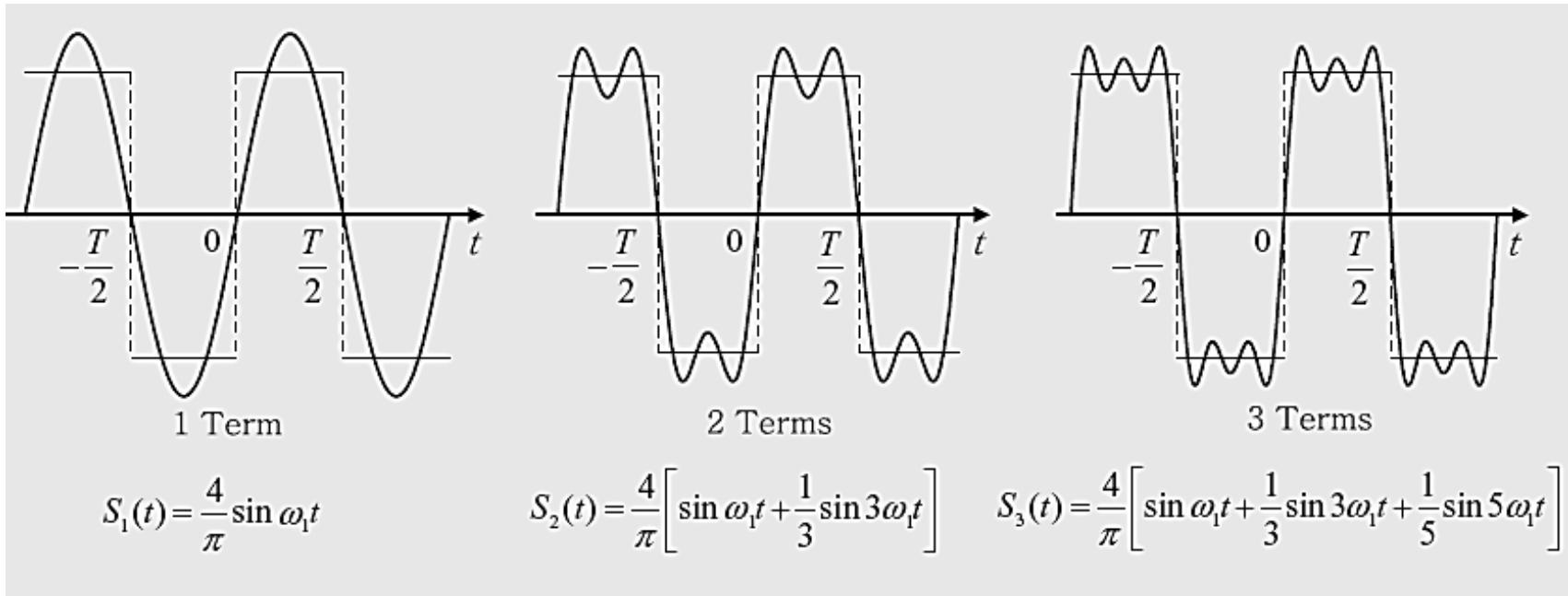
$$b_n = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} x(t) \sin\left(\frac{2n\pi t}{T}\right) dt = \frac{2}{n\pi} (1 - n\pi)$$

$$x(t) = \frac{4}{\pi} \left[\sin\left(\frac{2\pi t}{T}\right) + \frac{1}{3} \sin\left(\frac{2\pi 3t}{T}\right) + \frac{1}{5} \sin\left(\frac{2\pi 5t}{T}\right) + \dots \right]$$

$$x(t) = \frac{4}{\pi} \left[\sin \omega_1 t + \frac{1}{3} \sin 3\omega_1 t + \frac{1}{5} \sin 5\omega_1 t + \dots \right]$$



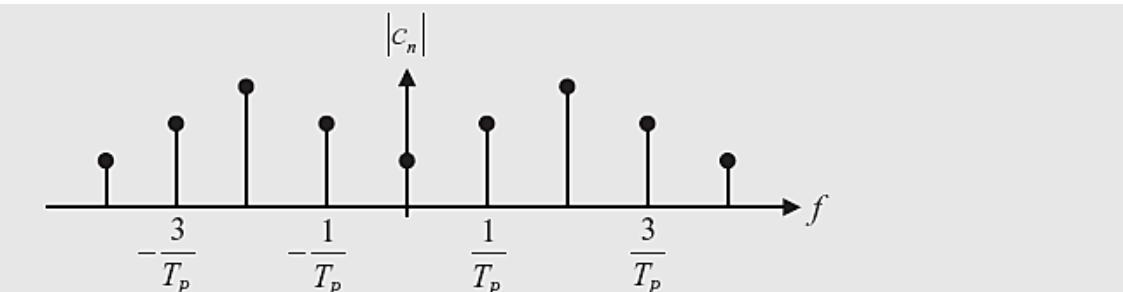
Fourier Series: Example



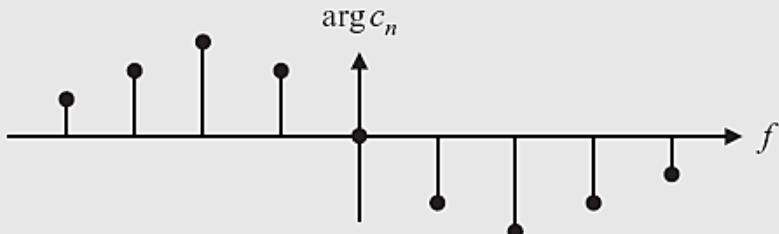
Amplitude & Phase Spectrum

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{\frac{j2\pi nt}{T}}$$

$$c_n = \frac{2}{T} \int_0^T x(t) e^{-\frac{j2\pi nt}{T}} dt$$



Amplitude spectrum of a Fourier series (a *line spectrum* and an *even function*)



Phase spectrum of a Fourier series (a *line spectrum* and an *odd function*)

MATLAB: example

```
clc; close all; clear all
```

% Keep hitting enter button if you want to see the term by term approx.

```
t=[0:0.001:1];  
x=[ ]; x_tmp=zeros(size(t));
```

```
for n=1:2:39  
    x_tmp=x_tmp+4/pi*(1/n*sin(2*pi*n*t));  
    x=[x; x_tmp];  
end
```

```
figure,
```

```
for i=1:20
```

```
drawnow
```

```
plot(t, x(i,:)) %plot(t,x(i,:),t,x(7,:),t,x(20,:));  
xlabel('itt\rm (seconds)'); ylabel('itx\rm(\itt\rm)')  
grid on
```

```
pause
```

```
end
```



FOURIER TRANSFORM

- Extension of Fourier analysis to non-periodic phenomena
- Discrete to continuous
- Skipping essential steps, in the limit $T_p \rightarrow \infty$

$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$$

← Fourier transform

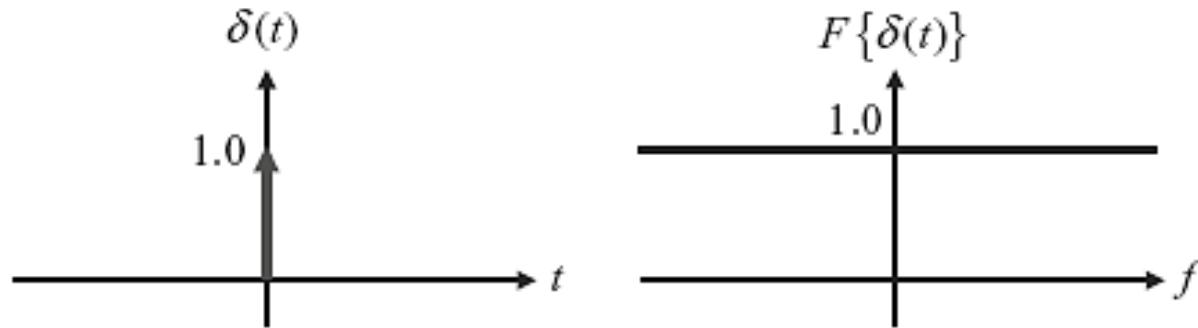
$$x(t) = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$$

← Inverse
Fourier transform



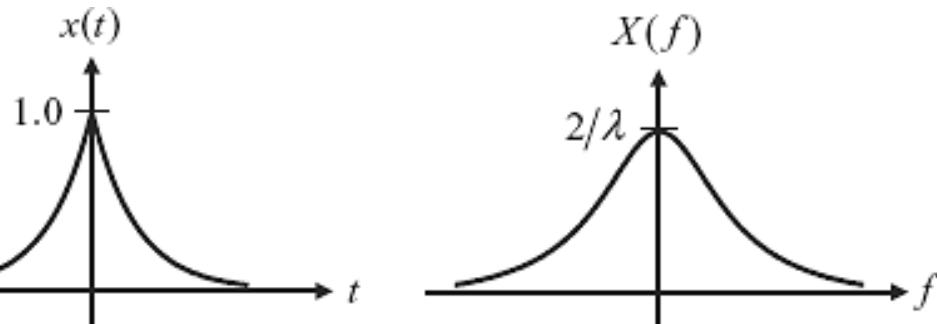
Examples

1) Dirac delta



2) Symmetric exponential

$$x(t) = e^{-\lambda|t|}, \quad \lambda > 0$$

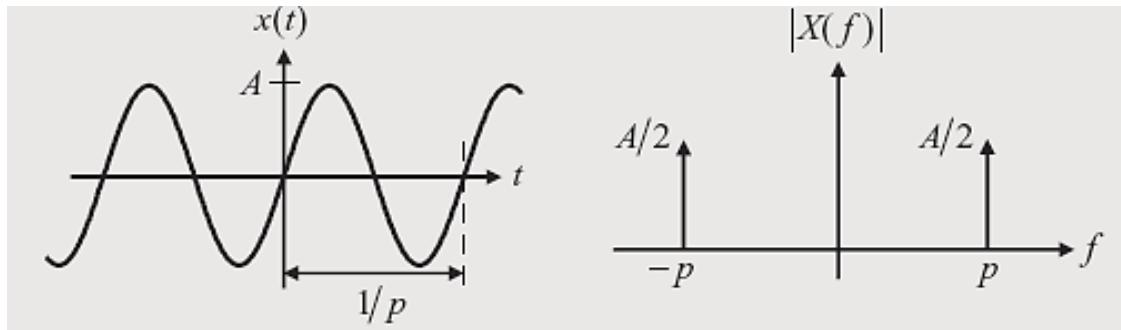


$$X(f) = \frac{2\lambda}{\lambda^2 + 4\pi^2 f^2}$$

Examples

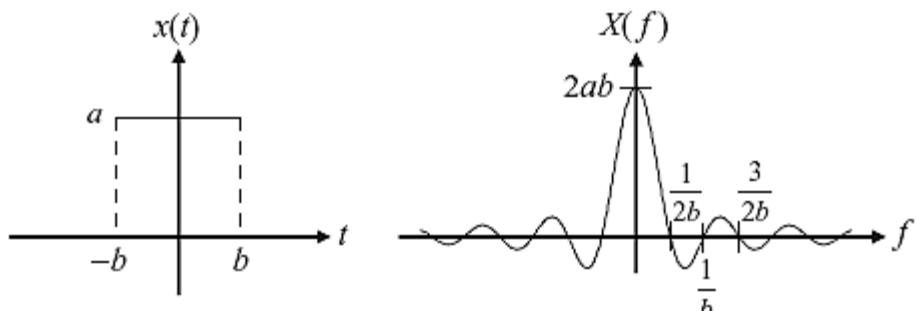
3) Sinusoid

$$x(t) = \sin(2\pi f_0 t) \text{ or } \sin(\omega_0 t) \quad X(f) = \frac{1}{2j} [\delta(f - f_0) - \delta(f + f_0)]$$



4) Window function

$$\begin{aligned} x(t) &= a & |t| < b \\ &= 0 & |t| > b \end{aligned}$$



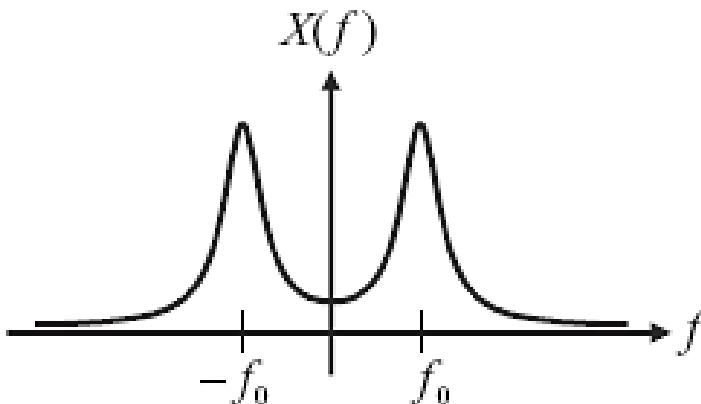
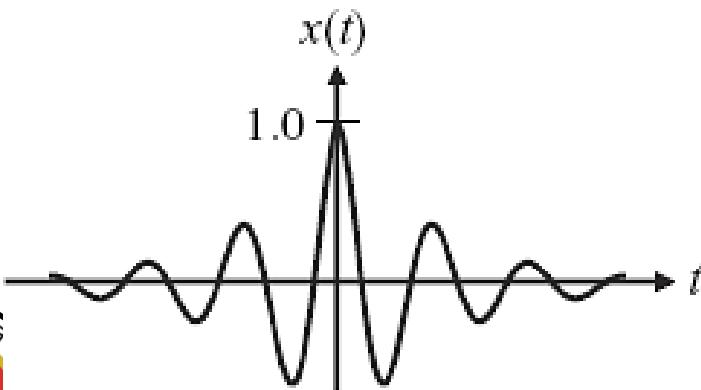
$$X(f) = \frac{2ab \sin(2\pi f b)}{2\pi f b}$$

Examples

5) Damped symmetrically oscillating function

$$x(t) = e^{-a|t|} \cos 2\pi f_0 t, \quad a > 0$$

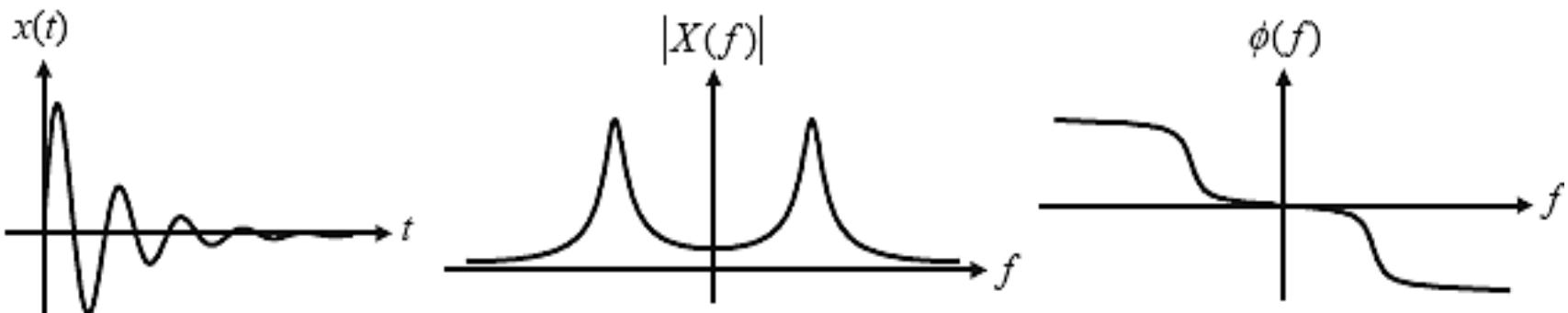
$$X(f) = \frac{a}{a^2 + [2\pi(f - f_0)]^2} + \frac{a}{a^2 + [2\pi(f + f_0)]^2}$$



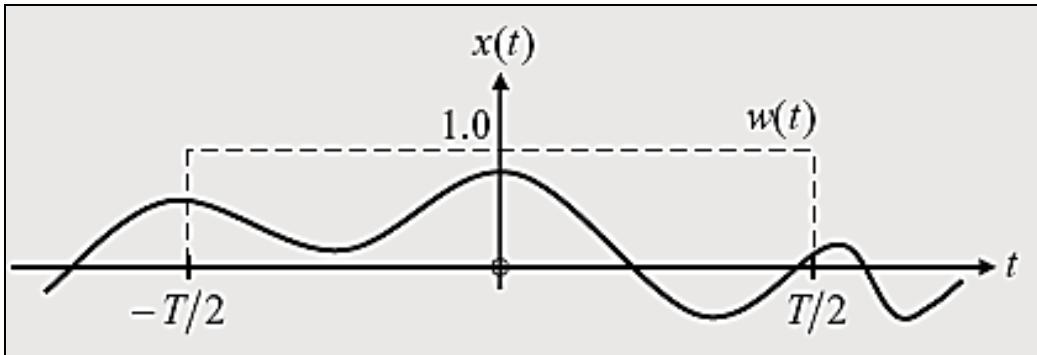
6) Damped oscillating function

$$x(t) = e^{-at} \sin 2\pi f_0 t, \quad t \geq 0 \text{ and } a > 0$$

$$X(f) = \frac{2\pi f_0}{(2\pi f_0)^2 + (a + j2\pi f)^2}$$



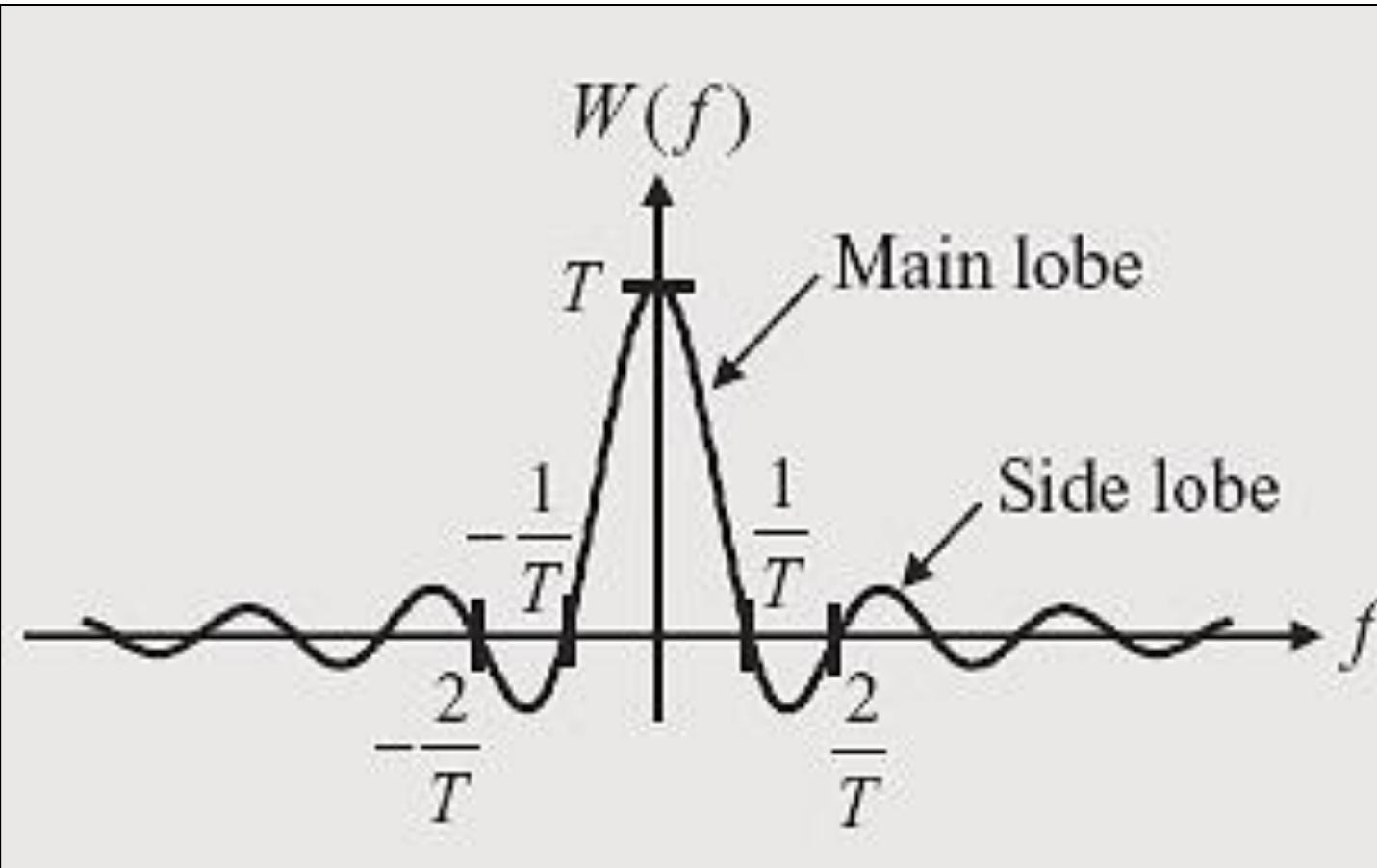
Windowing



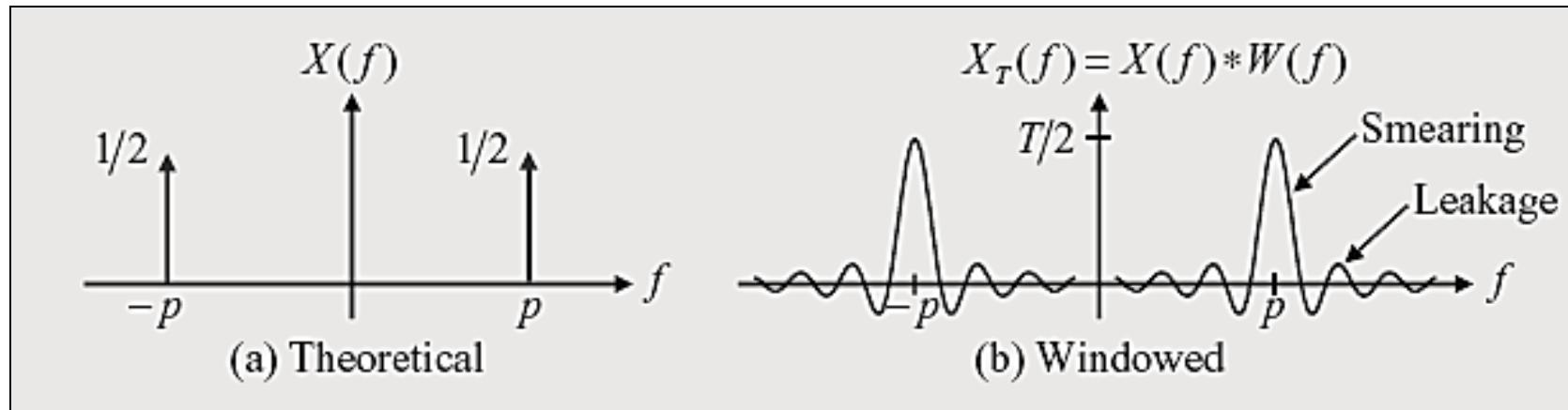
$$\begin{aligned}w(t) &= 1 & |t| &< T/2 \\&= 0 & |t| &> T/2\end{aligned}$$

$$\begin{aligned}X_T(f) &= \int_{-\infty}^{\infty} X(g)W(f-g)dg = \frac{1}{2} \int_{-\infty}^{\infty} [\delta(g+p) + \delta(g-p)] W(f-g)dg \\&= \frac{1}{2} [W(f+p) + W(f-p)]\end{aligned}$$

Windowing



Windowing: Illustration

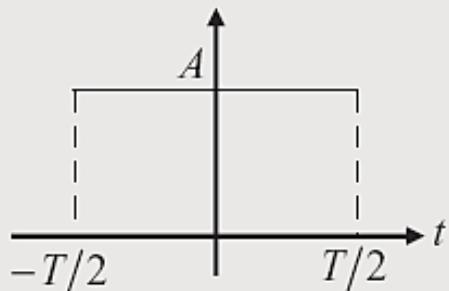


The distortion due to the main lobe is sometimes called *smearing*, and the distortion caused by the side lobes is called *leakage*.

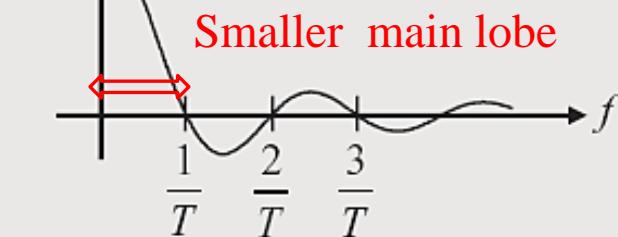
COMMON WINDOW FUNCTIONS

Rectangular Window

$$w(t) = A[u(t + T/2) - u(t - T/2)]$$

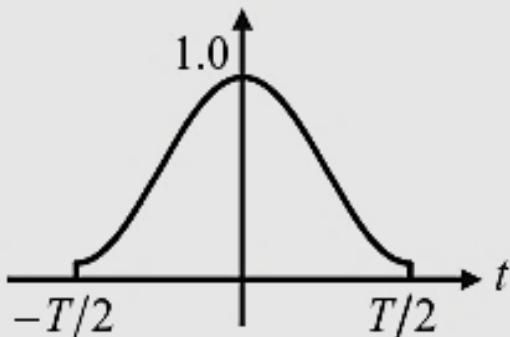


$$W(f) = AT \frac{\sin(\pi fT)}{\pi fT}$$

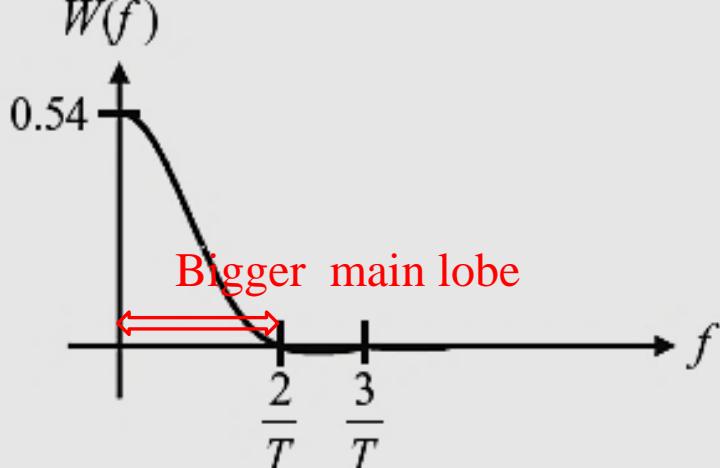


Hann Window

$$w(t)$$



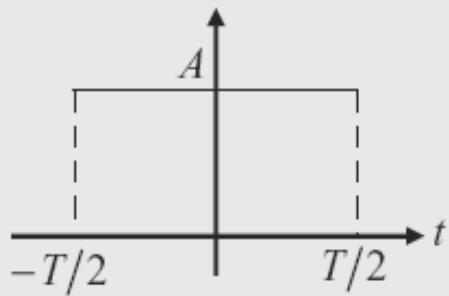
$$W(f)$$



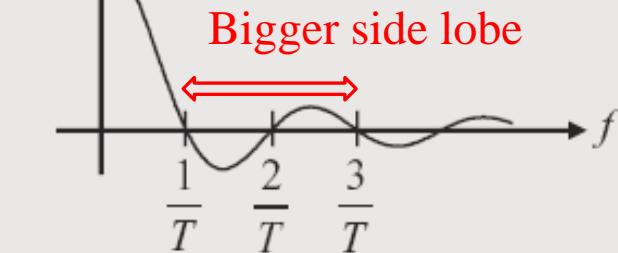
COMMON WINDOW FUNCTIONS

Rectangular Window

$$w(t) = A[u(t + T/2) - u(t - T/2)]$$

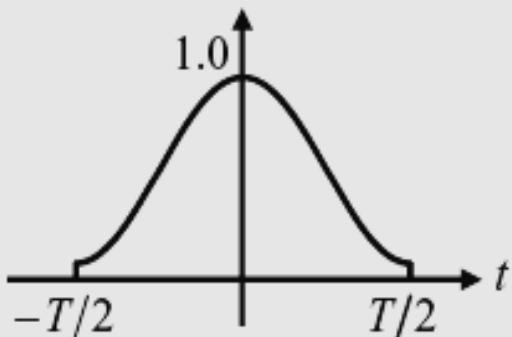


$$W(f) = AT \frac{\sin(\pi fT)}{\pi fT}$$

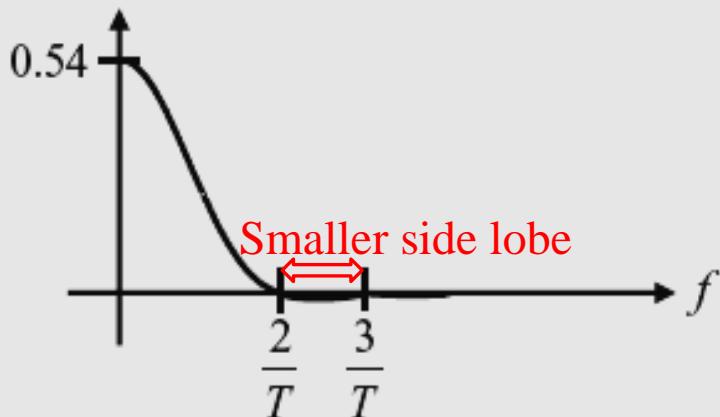


Hann Window

$$w(t)$$



$$W(f)$$



Discrete Fourier Transform

- Consider a sequence $x(n\Delta)$ at $n = 0, 1, 2, 3, 4, \dots, N-1$ points. The DFT is defined as :

$$X(e^{j2\pi f\Delta}) = \sum_{n=0}^{N-1} x(n\Delta)e^{-j2\pi fn\Delta}$$

- Note that this is still continuous in frequency



Discrete Fourier Transform

Now let us evaluate this at frequencies: $f = k/N\Delta$

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j(2\pi/N)nk}$$

$$X(k) = \left[X(e^{j2\pi f\Delta}) \text{ evaluated at } f = \frac{k}{N\Delta} \text{ Hz} \right] (k \text{ integer})$$

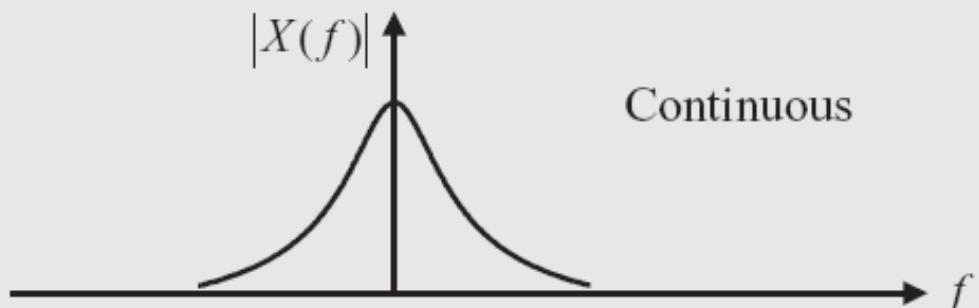
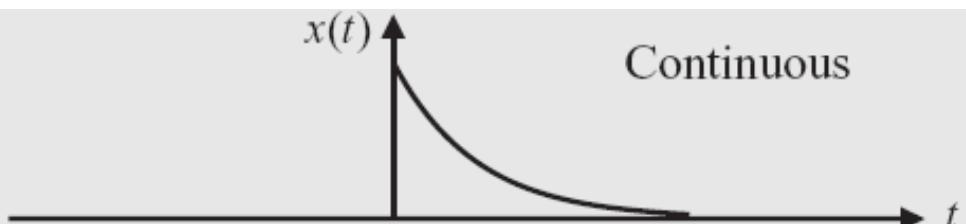


FOURIER INTEGRAL VS DFT

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df$$

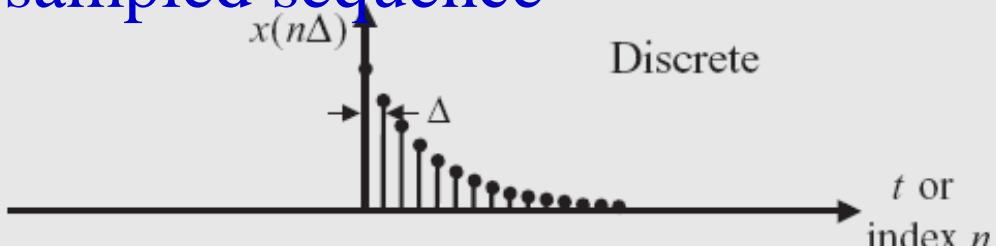
Fourier Integral

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

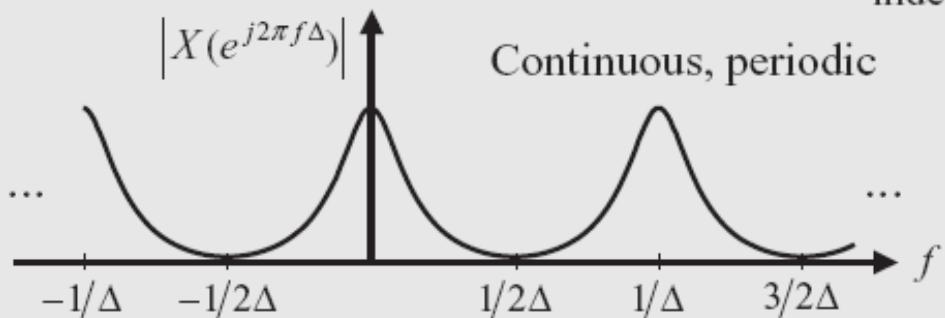


Fourier transform of the sampled sequence

$$x(n\Delta) = \Delta \int_{-1/2\Delta}^{1/2\Delta} X(e^{j2\pi f\Delta}) e^{j2\pi fn\Delta} df$$



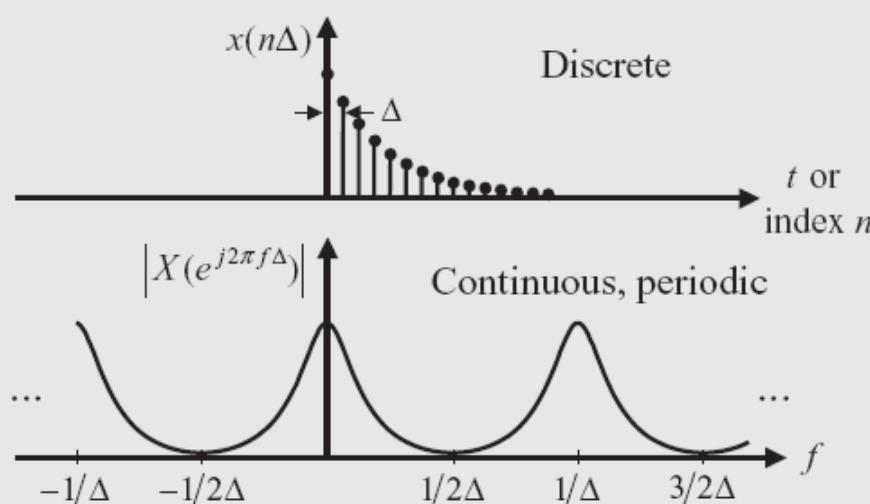
$$X(e^{j2\pi f\Delta}) = \sum_{n=-\infty}^{\infty} x(n\Delta) e^{-j2\pi fn\Delta}$$



FOURIER INTEGRAL VS DFT

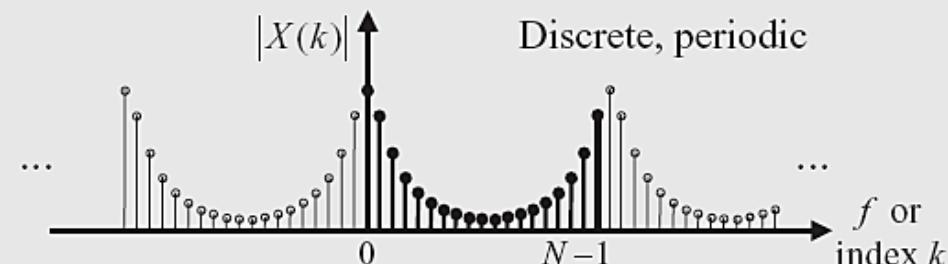
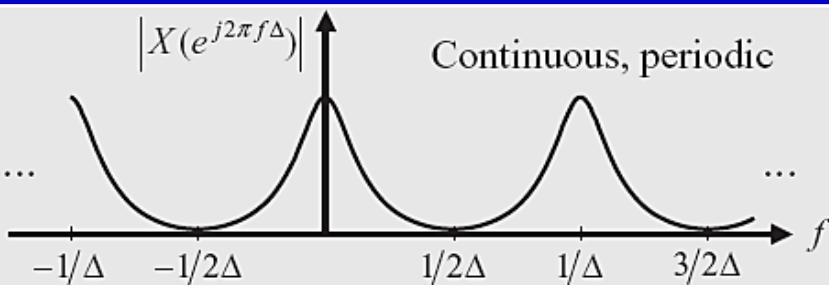
$$x(n\Delta) = \Delta \int_{-1/2\Delta}^{1/2\Delta} X(e^{j2\pi f\Delta}) e^{j2\pi fn\Delta} df$$

$$X(e^{j2\pi f\Delta}) = \sum_{n=-\infty}^{\infty} x(n\Delta) e^{-j2\pi fn\Delta}$$

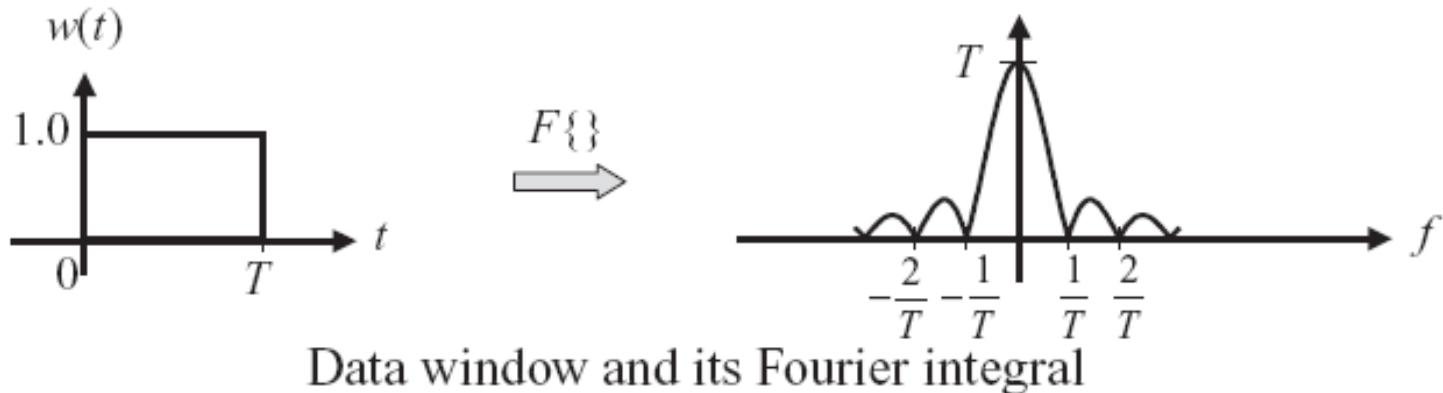
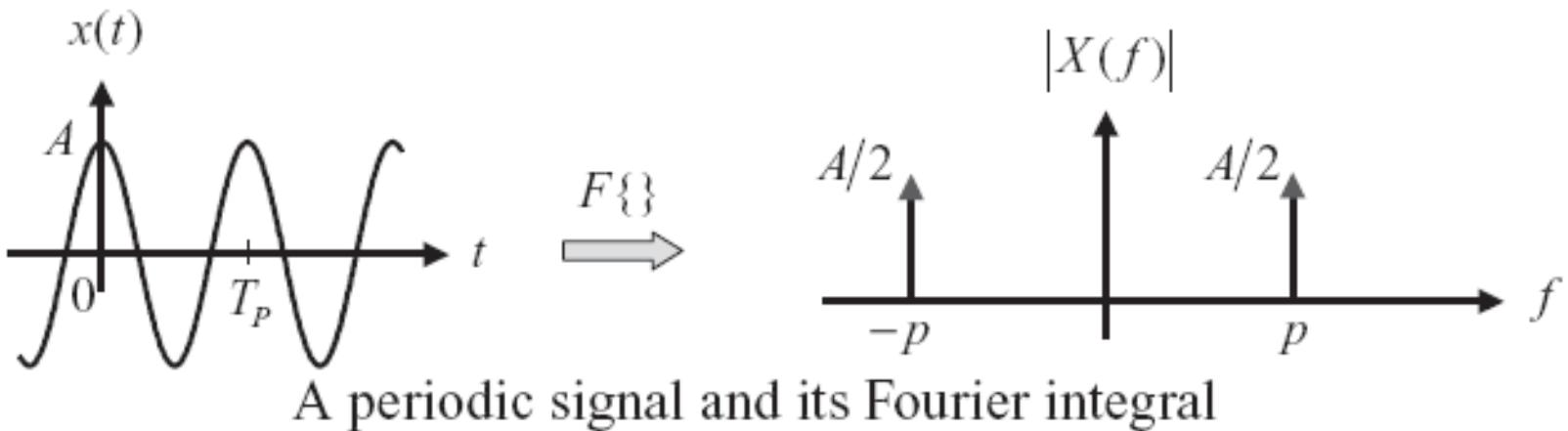


$$X(e^{j2\pi f\Delta}) = \sum_{n=-\infty}^{\infty} x(n\Delta) e^{-j2\pi fn\Delta}$$

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j(2\pi/N)nk}$$



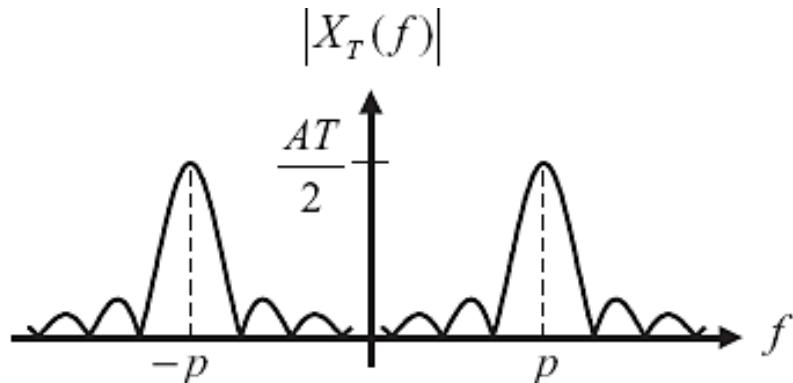
FOURIER INTEGRAL VS DFT



FOURIER INTEGRAL VS DFT

$$x_T(t) = w(t) \cdot x(t)$$

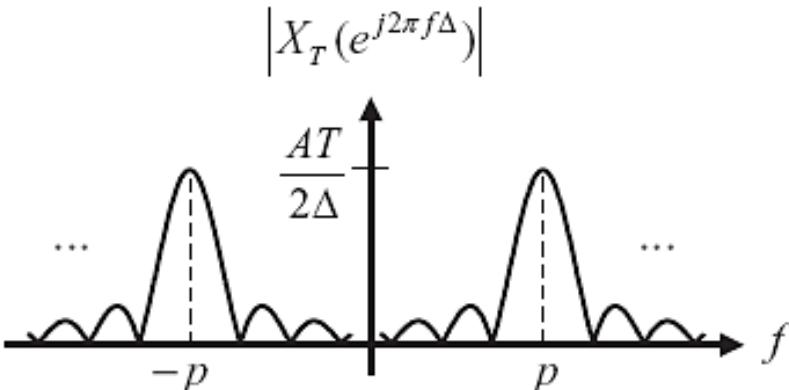
$\xrightarrow{F\{\cdot\}}$



(c) Truncated signal and its Fourier integral

$$x_T(n\Delta) = w(n\Delta) \cdot x(n\Delta)$$

$\xrightarrow{F\{\cdot\}}$



(d) Truncated and sampled signal and its Fourier transform of a sequence

FFT ALGORITHM: GLIMPSES

- The DFT provides uniformly spaced samples of the Discrete-Time Fourier Transform (DTFT)
- DFT definition:

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi n k}{N}}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j \frac{2\pi n k}{N}}$$

- Requires N^2 complex multiplications & $N(N-1)$ complex additions



LET'S TAKE AN EXAMPLE OF FFT: A SIMPLE MATLAB CODE

- Consider a signal: $x = A \sin 2\pi pt$
- $p = \frac{1}{T_p} \cdot \text{Hz}$
- Set a sampling frequency $fs = \frac{10}{T_p} \cdot \text{Hz}$
- $A = 2, p = 1 \text{ Hz}$
- What does Fourier integral give ? $|X(f)| = \frac{A}{2}$ at $p = 1 \text{ Hz}$

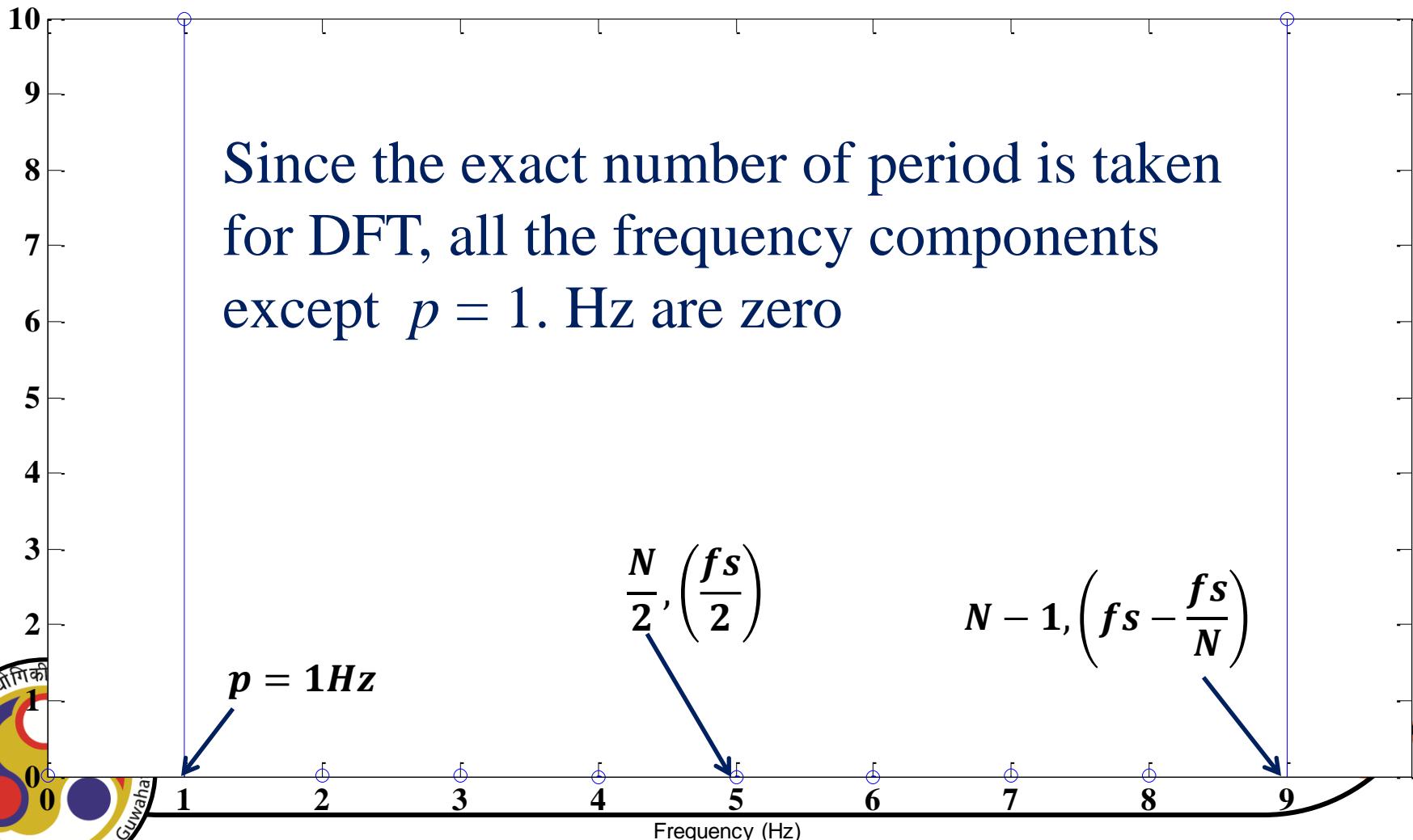


FFT MATLAB EXAMPLE-1

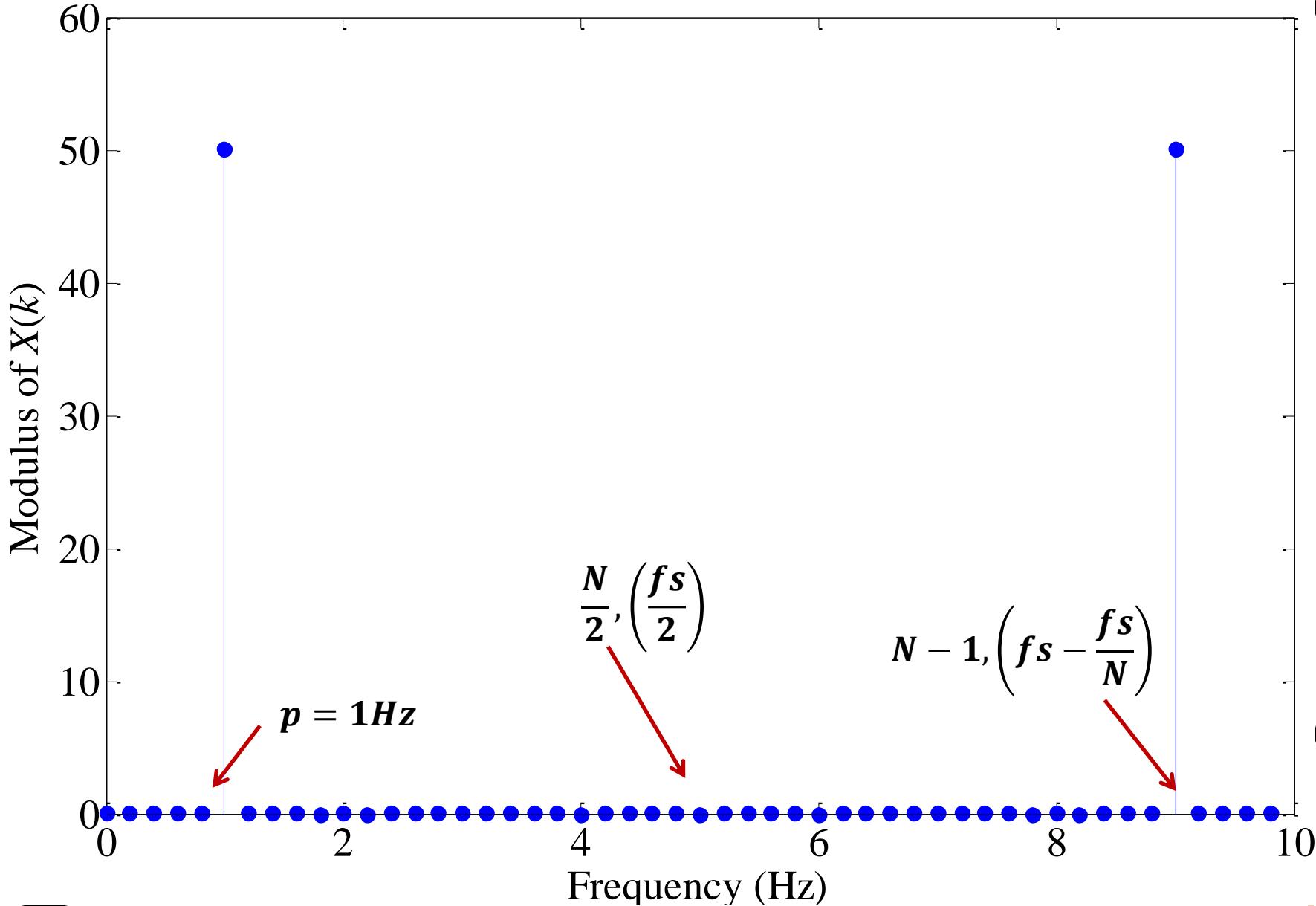
```
freq = 1; % Frequency of sinusoid. freq= 1/ Tp;  
  
Tp=1/freq; % Tp is the period  
  
fs=10*freq; % Define sample time and sample frequency. Take fs = 10/Tp  
  
ts=1/fs;  
  
T1= 1*Tp; % Time point where the data is truncated to  
  
t=0:ts:T1-ts;  
  
x= 2*cos(2*pi*freq*t);  
  
L=length(x);  
  
FT = fft(x); % FFT command; gives Fourier amplitude  
  
f =fs* (0:(L-1))/L; % Creating the X-axis or frequency axis  
  
fabs=abs(FT); % Single sided Fourier transform
```



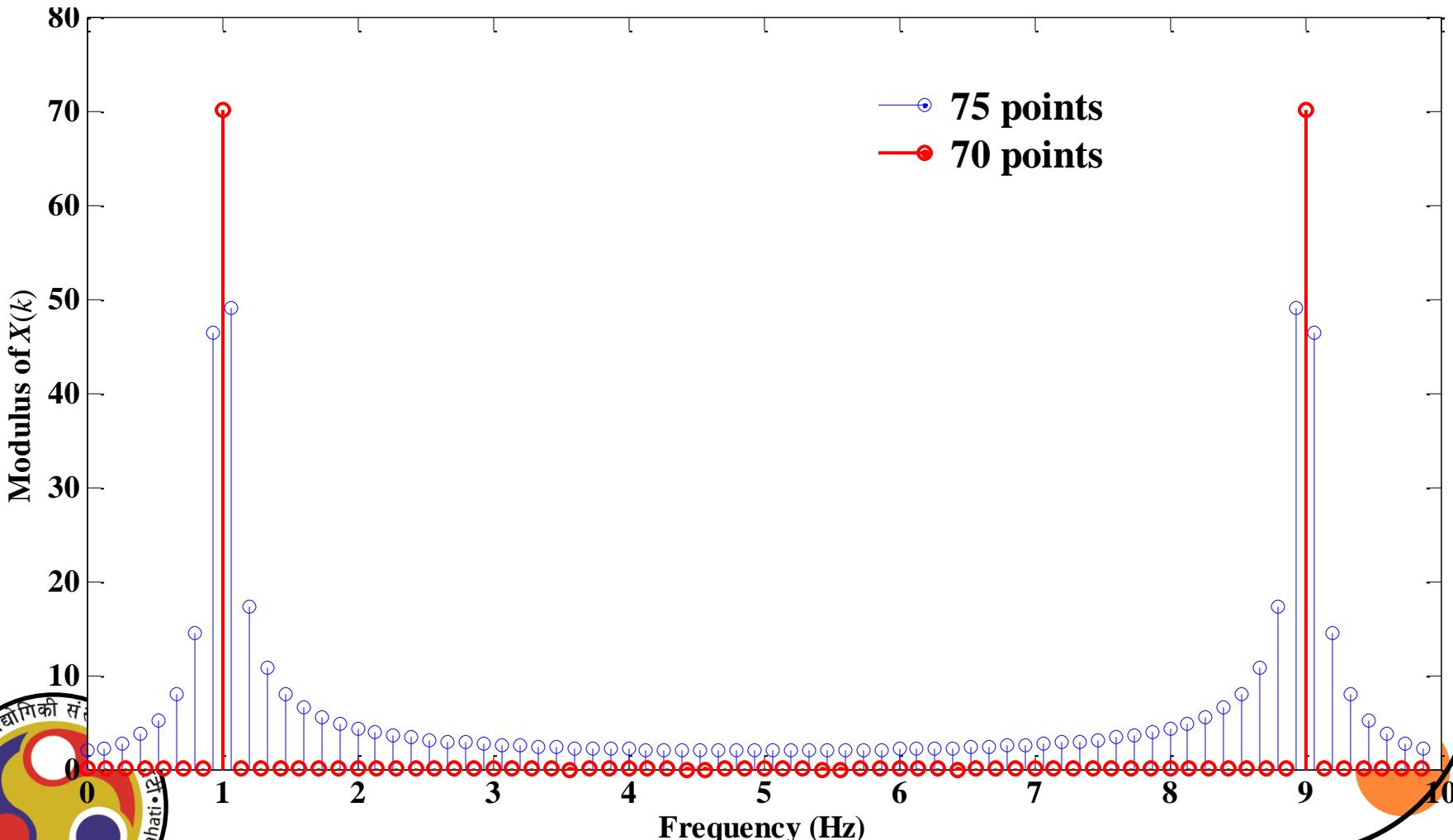
DATA TRUNCATED AT EXACTLY ONE PERIOD (10-POINT FFT)



DATA TRUNCATED AT 5 PERIODS (50-POINT FFT)



LEAKAGE



FFT MATLAB EXAMPLE-2

```
clc; clear all; close all  
A=2;  
p=1;  
Tp=1/p;  
fs=10/Tp;  
  
No_of_periods = 2; %Keep changing this parameter  
  
T1=No_of_periods*Tp;  
  
t1=[0:1/fs:T1-1/fs];  
  
x1=A*cos(2*pi*p*t1);  
  
X1=fft(x1);  
  
N1=length(x1); f1=fs*(0:N1-1)/N1;  
  
figure(1)  
subplot(1,2,1)  
plot(f1, abs(X1), 'b')  
xlabel('Frequency (Hz)')  
ylabel('Modulus of |itX|rm(|itk|rm)'); %axis([0 9.9 0 10])  
subplot(1,2,2)  
stem(f1, abs(X1), 'b')  
xlabel('Frequency (Hz)')  
ylabel('Modulus of |itX|rm(|itk|rm)'), %axis([0 9.9 0 10])
```



EXTRA

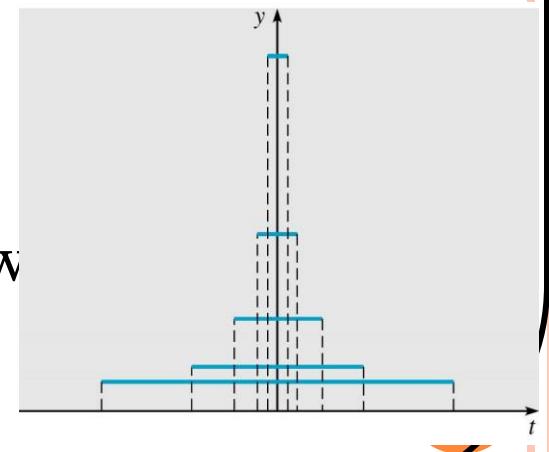
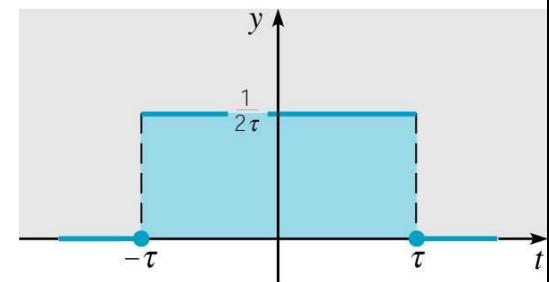


UNIT IMPULSE FUNCTION

- Suppose a function $d_\tau(t)$ has the form

$$d_\tau(t) = \begin{cases} 1/2\tau, & -\tau < t < \tau \\ 0, & \text{otherwise} \end{cases}$$

- Then, $I(\tau) = 1$.
- We are interested $d_\tau(t)$ acting over shorter and shorter time intervals (i.e., $\tau \rightarrow 0$). See graph on right.
- Note that $d_\tau(t)$ gets taller and narrower as $\tau \rightarrow 0$. Thus for $t \neq 0$, we have $\lim_{\tau \rightarrow 0} d_\tau(t) = 0$, and $\lim_{\tau \rightarrow 0} I(\tau) = 1$



DIRAC DELTA FUNCTION

- Thus for $t \neq 0$, we have $\lim_{\tau \rightarrow 0} d_\tau(t) = 0$, and $\lim_{\tau \rightarrow 0} I(\tau) = 1$
- The **unit impulse function** δ is defined to have the properties

$$\delta(t) = 0 \text{ for } t \neq 0, \text{ and } \int_{-\infty}^{\infty} \delta(t) dt = 1$$

- The unit impulse function is an example of a generalized function and is usually called the **Dirac delta function**.
- In general, for a unit impulse at an arbitrary point t_0 ,

$$\delta(t - t_0) = 0 \text{ for } t \neq t_0, \text{ and } \int_{-\infty}^{\infty} \delta(t - t_0) dt = 1$$



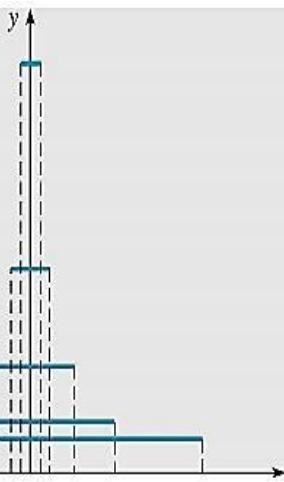
LAPLACE TRANSFORM OF δ

- The Laplace Transform of δ is defined by

$$L\{\delta(t-t_0)\} = \lim_{\tau \rightarrow 0} L\{d_\tau(t-t_0)\}, \quad t_0 > 0$$

and thus

$$L\{\delta(t-t_0)\} = \lim_{\tau \rightarrow 0} \int_0^\infty e^{-st} d_\tau(t-t_0) dt = \lim_{\tau \rightarrow 0} \frac{1}{2\tau} \int_{t_0-\tau}^{t_0+\tau} e^{-st} dt$$



$$\begin{aligned} &= \lim_{\tau \rightarrow 0} \frac{-e^{-st}}{2s\tau} \Big|_{t_0-\tau}^{t_0+\tau} = \lim_{\tau \rightarrow 0} \frac{1}{2s\tau} \left[-e^{-s(t_0+\tau)} + e^{-s(t_0-\tau)} \right] \\ &= \lim_{\tau \rightarrow 0} \frac{e^{-st_0}}{s\tau} \left[\frac{e^{s\tau} - e^{-s\tau}}{2} \right] = e^{-st_0} \left[\lim_{\tau \rightarrow 0} \frac{\sinh(s\tau)}{s\tau} \right] \\ &= e^{-st_0} \left[\lim_{\tau \rightarrow 0} \frac{s \cosh(s\tau)}{s} \right] = e^{-st_0} \end{aligned}$$

LAPLACE TRANSFORM OF δ

- Thus the Laplace Transform of δ is

$$L\{\delta(t-t_0)\} = e^{-st_0}, \quad t_0 > 0$$

- For Laplace Transform of δ at $t_0=0$, take limit as follows:

$$L\{\delta(t)\} = \lim_{t_0 \rightarrow 0} L\{d_\tau(t-t_0)\} = \lim_{\tau_0 \rightarrow 0} e^{-st_0} = 1$$

- For example, when $t_0=10$, we have $L\{\delta(t-10)\} = e^{-10s}$.

